High Electro-Optic Sensitivity ($r_{33}$) Polymers: They Are Not Just for Low Voltage Modulators Any More†

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Received: September 24, 2003; In Final Form: April 19, 2004

To achieve gain ≥0 dB in an external modulation analogue optical link with modest laser power (~10 mW), the external modulator needs to have an on–off voltage ($V_o$) of ~ 0.3 V, which is more than a factor of 10 smaller than the on–off voltages of most commercially available modulators. Polymeric materials, in which the electrooptic tensor $r_{33}$ has been engineered to have a very large magnitude (> 100 pm/V), enable external modulator designers to meet this goal, because the modulator’s on–off voltage is inversely proportional to this tensor magnitude. Now that polymer materials have surpassed 100 pm/V, the natural question is: what do we need even higher $r_{33}$ material for? We will show that there are many uses to which a larger $r_{33}$ material can be put, but that, contrary to present perception, even lower $V_o$ is not one of them. The paper concludes by discussing one of the uses for a larger $r_{33}$: a linearized modulator.

Introduction

Fiber optic links have become ubiquitous for long distance digital communication and are finding increasing application for transporting analogue signals as well. Regardless of whether one is analyzing a distribution link that feeds hundreds or thousands of nodes, such as with cable TV (CATV), all of these link configurations can be decomposed into one or more basic point-to-point optical link(s), as shown in Figure 1a. For the RF-to-optical modulation device, both direct and external modulation have been extensively investigated and are in widespread commercial use. Since the discussion in this paper will center on electrooptic polymers for external modulators, we will limit our consideration to such external modulators in general and to the Mach–Zehnder modulator in particular. The layout of this modulator is shown in Figure 1b and is one of the most common forms of such modulators. The modulating voltage is applied to electrodes that are placed above single mode optical waveguides, which have been fabricated in an electrooptic material. The resulting electrical field alters the optical index of refraction in the waveguide to an extent that is largely determined by the material’s electrooptic tensor, which is commonly designated by $r_{33}$. The interferometric combination of two phase modulated optical waves results in an intensity modulated optical wave that corresponds to the original electrical modulation signal.

One of the key parameters of an optical link is its RF loss, which is defined simply as the ratio of the RF power at the link output to the RF power at the link input. In turn it can be shown (see for example ref 1) that one of the most effective techniques for reducing the loss, or even achieving RF gain, from a link is to have a modulator with a low switching voltage. In the case of a Mach–Zehnder modulator, this would mean a low $V_o$, which is the voltage required to switch the modulator between full on and full off.

In addition to $r_{33}$, $V_o$ also depends on several geometrical parameters involving the electrodes and their layout relative to the optical waveguide. These parameters are defined in Figure 1b. An expression for the $V_o$ of a Mach–Zehnder modulator can be show to be

$$V_o = \frac{\lambda g}{pn |r_{33}| \Gamma L}$$

where $g$ and $L$ are as defined in Figure 1b, $\lambda$ is the optical wavelength, $p$ is a constant that depends on electrode geometry, $\Gamma$ is the overlap of the modulating and optical fields, and $n$ is the electrooptic material refractive index. This equation makes clear that once all of the other parameters have been optimized the only remaining option to achieve a low $V_o$ is as large a value of $r_{33}$ as possible. In theory, one might also try reducing $V_o$ by increasing $L$, but electrical loss will dictate diminishing returns, and large $L$ is also undesirable from a packaging standpoint.

$V_o$: Low but Not Too Low

We begin our investigation of the impacts of lower $V_o$ on link performance by considering the link gain. In Figure 2, we plot the intrinsic link gain (i.e., the link gain without any amplifiers) vs the $V_o$ of the Mach–Zehnder modulator, with

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† Part of the special issue “Alvin L. Kwiram Festschrift.”

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of link gain, which in turn are more effective at reducing the impact of noise at the link output on the noise figure. However, eventually, we reach a region where further decreases in $V_\pi$ do not result in further decreases in noise figure. In this region, the input noise is amplified by the high-gain link to such a degree that other noise generated by the link is comparatively insignificant. Notice that this constant region has gone largely un-noticed at present because the range of $V_\pi$ values for current Mach–Zehnder modulators is above the region where one would expect to see this effect. The “knee” value of $V_\pi$ that signifies the transition between these two regions depends on the average optical power.

A third important link parameter is dynamic range. There are actually several measures of dynamic range. The one we will concentrate on here is the intermodulation-free dynamic range (IMFDR), which is the ratio of the fundamental to distortion powers, when the distortion signal power equals the noise power. The power of the RF signal applied to a modulator at which the power of the output distortion products equals that of the noise depends on the linearity of the modulator, as will be discussed further on in this paper. Basically, the more linear the modulator, the greater the power the input RF signal can have before distortion products exceed the noise and, therefore, the greater the IMFDR. We plot in Figure 3 the IMFDR vs Mach–Zehnder for the same range of optical powers. For high values of $V_\pi$, the IMFDR is independent of $V_\pi$. The reason is that in this range the distortion is changing at the same rate as the noise; hence, their ratio is independent of $V_\pi$. However, once one enters the region where the noise figure ceases to decrease with $V_\pi$, then the decrease in IMFDR with further decreases in $V_\pi$ becomes apparent. Again this region is largely unexplored, because it is below the $V_\pi$ range of present commercial Mach–Zehnder modulators.

Now by examining both noise figure and IMFDR in Figure 3, it is apparent that $V_\pi$ introduces a link design tradeoff. Unlike the link gain dependence on $V_\pi$, where smaller is always better, we see that when these two additional link parameters are taken into account small values $V_\pi$ of are not always better. Often one wants the best of both noise figure and IMFDR. To satisfy this requirement, one would want a small enough value of $V_\pi$ that the noise figure is low, while at the same time remaining on the edge of the constant IMFDR region. To satisfy this tradeoff, we clearly see that the smallest value of $V_\pi$ is not the best value.

To get a feel for the values of $r_{33}$ where this tradeoff comes into effect, we have summarized in Table 1 the knee values of $r_{33}$ as a function of $V_\pi$ and laser optical power into the Mach–Zehnder modulator, for the typical modulator design parameters listed in the Figure 1 caption. Although the range of 360–3600 pm/V may seem large to some, it is important to point out that it is projected that within the next few years electrooptic polymers will at least reach the lower end of this range. (For a discussion of the chemical routes to these high $r_{33}$ values, see ref 4 and the articles referenced therein.)

**Beyond Low $V_\pi$: A Use for Even Higher Values Of $r_{33}$**

It might appear from the preceding section that there is not much utility to attempting to achieve further increases in $r_{33}$.
However, that is not the case if one considers applications of modulators that go beyond the simple, point-to-point link example that was discussed above. Some of the examples of where higher values of $r_{33}$ could be put to good use include:

1. Linearized modulators
2. Polarization insensitive modulators
3. Frequency conversion links

We will discuss only the first example in the following. It turns out that there are a number of applications, antenna remoting of radar signals and distribution of CATV signals prime among them, where greater IMFDR is required than is available from a standard Mach–Zehnder modulator. As mentioned earlier in this paper, the upper end of the IMFDR is dictated by the linearity of the processes by which the RF signal at the input to an analogue link modulates the light and is retrieved from its optical carrier at the detector at the output of the link. In most analogue links, the modulation process is quite nonlinear (especially when compared to the most commonly used detection process, which is very linear over a large range of optical and RF power levels). For example, the depth of optical modulation from a Mach–Zehnder modulator is proportional to the trigonometric sine of a quantity that is proportional to the input signal. For small signals, the small-angle approximation of the sine function tells us that the modulation is quite linear, and therefore, we can expect distortion products to be weak (below noise). However, for larger input signals, the small-angle approximation no longer holds, and the nonlinearity of the modulation results in strong distortion products. Hence, over the years, a number of methods have been proposed, investigated, and commercialized that extend the link IMFDR via “linearization” of the modulation function. An example of one configuration of linearized Mach–Zehnder modulator is shown in Figure 4 (after ref 5). In this linearization method, two Mach–Zehnder modulators are connected in parallel. Both are modulated with the same signal, albeit with different amplitudes. The bias points of the modulators are chosen such that in combination with the modulation amplitudes the distortion produced by each modulator is of equal amplitude but 180° out of phase from the other. Thus, the distortion is canceled.

Unfortunately, the modulation at the fundamental is partially, but not completely, canceled as well. Thus, while linearization eliminates (at least ideally) the distortion, it also reduces the fundamental. Although we have shown this for this particular linearization topology, it turns out that this effect occurs in all other optical linearization methods of which the authors are aware.

If we now repeat Figure 4 for the case of a linearized modulator, we obtain the plot shown in Figure 5. Although the general shape of the noise figure and IMFDR curves is the same as in the previous plot, we notice that the “knee” values have been shifted to lower values of $V_n$. In turn, this means that to achieve the same tradeoff point with a linearized modulator is going to require a smaller value of $V_n$, which in turn will require a higher value of $r_{33}$.

Conclusions

We have tried to demonstrate that a higher value of electrooptic material $r_{33}$ has many benefits, prime among them are higher link gain and lower noise figure. However, we have also demonstrated that if an additional link parameter, IMFDR, is also included, as it almost invariably is, that the link improvements with increasing $r_{33}$ only go so far. Beyond this value, further increases in $r_{33}$ are actually detrimental to link performance. Hence, a tradeoff exists and the exact value depends on the modulator design parameters as well as the average optical power that will be fed to the modulator.

However, there are definitely uses of higher values of $r_{33}$. There are many link and modulator configurations in which an improvement in one parameter comes at the expense of another. In these cases higher values of $r_{33}$ would permit one to maintain the improved parameter while simultaneously reducing the degradation of the other. The linearized modulator is but one example of this; other examples include a polarization independent modulator and a frequency converting link.

Acknowledgment. The authors thank Professor Larry Dalton for inviting them to present this work and for many stimulating discussions. The authors also thank Professor Bill Bridges of Caltech for many stimulating discussions as well and Charles Lee of AFOSR for supporting the work.

References and Notes